

## SOLID STATE IMAGE PICKUP APPARATUS

### BACKGROUND OF THE INVENTION

#### Field of the Invention

5           The present invention relates to a solid state image pickup apparatus, and more particularly to a solid state image pickup apparatus having a floating diffusion unit adjacent to the photoelectric conversion unit.

#### 10       Related Background Art

          Conventionally, the charge coupled device (CCD) is widely utilized as the solid state image pickup apparatus. Fig. 1 is a schematic cross-sectional view of a common CCD, and Fig. 2 shows the change of the potential distribution in time of various areas shown in Fig. 1, wherein areas same as those in Fig. 1 are represented by same numbers.

          In Fig. 1 there are shown a Si semiconductor substrate 1 of a first conductive type, a well area 2 of a second conductive type opposite to that of the substrate 1, and a diffusion area 3 of the first conductive type, constituting a PN photodiode with the well area 2. The PN photodiode is initially given an inverse bias and a photo-induced charge is accumulated according to the incident light amount, in the junction capacitance of the PN photodiode. There are also shown a shallow diffusion area 4 of the second conductive

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type, provided for reducing the dark current generated on the surface of the diffusion area 3 and forming a diode also between the areas 3 and 4 thereby increasing the amount of the accumulated charge, a transfer channel 5 for reading the photo-induced charge accumulated in the photodiode, and a CCD channel 6 for successive transfer of the read charge.

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10 A control electrode 7 serves to control the charge transfer to the CCD and also the successive charge transfer in the CCD, and is formed on the transfer channel 5 and the CCD channel 6 across an insulation layer. Second, third and fourth electrodes 8, 9, 10 serve to transfer the charge in the CCD in succession. A floating diffusion area 11 finally receives the  
15 transferred photo-induced charge, and a voltage amplitude, generated in the floating diffusion area 11 according to the photo-induced charge amount, is detected by a source follower amplifier, omitted in Fig. 1, and outputted to the outside as an electrical  
20 signal. There are also shown a reset gate 12 for resetting the floating diffusion area 11, and a reset drain 13 therefor.

25 In the following the functions will be explained with reference to Fig. 2, which illustrates the potential states of the various areas. At a time  $t_0$ , the transfer channel 5 is turned off and the photodiode area 2, 3 accumulates the photo-induced charge

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corresponding to the incident light amount. At a time  $t_1$ , the transfer channel is turned on to transfer the photo-induced charge to the CCD channel 6 through the transfer channel 5.

At times  $t_2$ ,  $t_3$ , the control electrodes 8, 9 are controlled in succession to transfer the photo-induced charge is transferred in succession to the portions in the CCD channel, positioned directly below the control electrodes. The floating diffusion area 11 is reset in advance at the time  $t_3$ , and, the photo-induced charge is transferred by the control electrode 10, at a time  $t_4$ , to the floating diffusion area 11 and is converted therein into a voltage for output.

In the above-described configuration, however, as the common control electrode 7 is used for controlling the potentials of the transfer channel 5 and the first CCD channel 6, the difference in the depth of the potential wells thereof is uniquely determined by the process conditions such as the concentration profile.

For this reason, there may result a case where the depth of the potential well of the CCD channel 6 cannot be sufficiently secured as shown in Fig. 3 in which areas same as those in Fig. 1 are represented by same numbers. As a result, the photo-induced charge cannot be fully transferred but partly remains in the transfer channel 5, leading to defective phenomena such as:

- (1) a decrease in the total amount of the

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transferred charge, leading to a lowered sensitivity;  
and

(2) return of the charge remaining in the  
transfer channel 5 to the photodiode 2, 3, thereby  
5 generating a retentive image and significantly  
deteriorating the image quality.

Also the Japanese Patent Application Laid-open No.  
2-30189 discloses a configuration having an electrode  
for controlling the charge transfer from the photodiode  
10 to the CCD channel and another electrode for  
controlling the charge transfer in the CCD, as shown in  
Fig. 5. Referring to Fig. 5, a plurality of n-type  
accumulation areas 1 for accumulating the signal  
charges induced by the incident light are arranged on a  
15 p-type substrate 6, and a transfer gate 13 is provided  
between the accumulation area 1 and a vertical CCD  
register 2, of which an end is connected to a  
horizontal CCD register composed of a transfer gate 12,  
a transfer electrode 13 for the vertical CCD register  
20 and a silicon dioxide layer 14.  $V_A$  denotes the  
potential of the n-type accumulation area 1, taking the  
Fermi potential 15 of the interior of the p-type  
substrate as a reference, while  $V_{DEP}$  denotes the  
potential  $V_A$  required for completely depleting the  
25 accumulation area 1, and  $V_{ch}$  denotes the channel  
potential when the transfer gate 3 is turned on. The  
potential  $V_A$  is lowered in the drawing in response to

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the light entering the accumulation area 1, and the accumulated charge is transferred to the vertical CCD register 2 by lowering the potential of the gate 12.

However, according to the above-mentioned patent application, the depth of the potential well of the CCD channel is still uniquely determined by the process conditions such as the concentration profile, so that there are similarly encountered the above-mentioned drawbacks.

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In the field of solid state image pickup apparatus, in addition to CCD, the MOS sensor is recently attracting attention and is actively developed because of the advantages such as ease of one-chip formation of the peripheral circuits. For example, an exhibit titled "An Active Pixel Sensor Fabricated Using CMOS/CCD Process Technology" (exhibited at 95 IEEE WORKSHOP on Charge-coupled Device and Advanced Image Sensors) discloses a CMOS sensor and its potential chart as shown in Fig. 4. As shown in Fig. 4, in the CMOS sensor, the floating diffusion area and the source follower amplifier are provided for each pixel, instead of at the end of the CCD register in case of the CCD. In Fig. 4, components same as those in Fig. 1 are represented by same numbers. There are also shown a transfer gate 701 provided for each pixel, and a floating diffusion area 711 provided for each pixel. The photo-induced charge, generated in the photodiode

area is transferred to the floating diffusion area 711, and the amplitude of the voltage generated therein according to amount of photo-induced charges is detected by the source follower amplifier, omitted in Fig. 4 but provided for each pixel, and outputted to an output line through a pixel selecting switch. There are also shown a reset gate 712 for resetting the floating diffusion area 711 provided for each pixel, and a reset drain 713 therefor.

10        Such circuit configuration of transferring the charge of each pixel to the corresponding floating diffusion area 711 and effecting the selection of pixels by a common MOS circuit allows achieve both a high S/N ratio as the sensor and a high performance realized by a one-chip MOS circuit.

15        However, such conventional configuration is still associated with the drawbacks similar to those in the conventional CCD configuration, because of changes in the reset voltage in the floating diffusion area and in  
20        the depth of the potential well of the transfer channel, caused by a fluctuation in the manufacturing process conditions. These drawbacks have not been paid any attention, and the potential chart of the  
25        aforementioned exhibit only showed a configuration that is incapable of complete transfer of the charge from the photodiode to the floating diffusion area.

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# SUMMARY OF THE INVENTION

The object of the present invention is to provide a configuration of the solid state image pickup apparatus, capable of complete transfer of the charge from the photoelectric conversion element to the floating diffusion area.

The above-mentioned object can be attained, according to a scope of the present invention, by providing a solid state image pickup apparatus comprising a photoelectric conversion element, transfer switch means consisting of a MOS transistor for transferring a signal charge generated in the photoelectric conversion element, a floating diffusion capacitance for receiving the signal charge through the transfer switch means, and reset switch means consisting of a MOS transistor for resetting the potential of the floating diffusion capacitance, wherein the pickup apparatus further comprises at least one potential setting means for generating a voltage different from the power supply voltage, and the output of the potential setting means is supplied as a pulse to the gate of the transfer switch means and/or the gate of the reset switch means, to control the gate potentials at the respectively optimum potentials.

According to an another aspect of the present invention, there is provided a solid state image pickup apparatus comprising a PN junction unit capable of

photoelectric conversion, a transferring MOS transistor for transferring a signal charge generated in the PN junction unit, a floating diffusion unit for receiving the signal charge through the transferring MOS transistor, a reset MOS transistor for resetting the potential of the floating diffusion unit, first potential setting means for supplying a voltage to be applied to the gate of the transferring MOS transistor and a second potential setting means for supplying a voltage to be applied to the resetting MOS transistor.

These configurations allows to obtain a satisfactory image.

Other objects of the present invention, and the features thereof, will become fully apparent from the following description, which is to be taken in conjunction with the attached drawings.

#### BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is a schematic cross-sectional view of a conventional CCD;

Fig. 2 is a chart showing the function of the conventional configuration;

Fig. 3 is a view showing a drawback in the conventional configuration;

Fig. 4 is a schematic cross-sectional view of a second configuration;

Fig. 5 is a view schematically showing first and

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second embodiments of the present invention and the function thereof;

Figs. 6A, 6B, 7A, 7B, 8 and 9 are views showing the function of the first embodiment of the present invention;

Figs. 10 and 11 are views showing the function of the second embodiment of the present invention; and

Figs. 12 and 13 are views showing the function of the first and second embodiments of the present invention.

#### DESCRIPTION OF THE PREFERRED EMBODIMENTS

Now there will be given a detailed description on a first embodiment of the present invention, with reference to the attached drawings. Fig. 6A is a schematic view showing the first embodiment of the present invention, wherein shown are a Si semiconductor substrate 1 of a first conductive type, a well area of a second conductive type opposite to that of the substrate 1, a diffusion area 3 of the first conductive type, forming a PN photodiode with the well area 2, a shallow diffusion area 4 of the second conductive type, a transfer channel 5 for reading the photo-induced charge accumulated in the photodiode, a floating diffusion area 11, a reset gate 12 for resetting the floating diffusion area 11, and a reset drain 13 therefor. Components same as those in Fig. 5 are

represented by same numbers and will not be explained further.

There are also provided a transfer gate 501 of a transfer switch for transferring the photo-induced charge in the photodiode 2, 3, 4, a control pulse generation circuit 504, a first potential setting circuit 502 for setting the voltage of the control signal generated by the control pulse generation circuit 504 for application to the transfer gate 501, and a second potential setting circuit 503 for setting the voltage of the control signal generated by the control pulse generation circuit 504 to supply that voltage to the reset gate 12.

The first and second potential setting circuits 502, 503 are so constructed that the transfer gate 501 and the reset gate 12 can be set at respectively independent potentials.

Now, the function of the present embodiment will be explained with reference to Fig. 6B showing the potential state of various areas at different timings. At a time  $t_0$ , the transfer channel 5 is turned off and a photo-induced charge is accumulated in the photodiode area 2 - 4 according to the amount of the incident light. The floating diffusion area 11 is reset by turning on the reset gate at a time  $t_1$ . Then, at a time  $t_2$ , the photo-induced charge is transferred from the photodiode area 2 - 4 through the transfer channel

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5 to the floating diffusion area 11 and converted therein into a voltage and outputted. At a time  $t_3$ , the transfer channel 5 is turned off again, and the photodiode area 2 - 4 of each pixel starts next charge accumulation.

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10 In these operations, as shown at time  $t_2'$  in Fig. 7A, the charge may remain in the transfer channel 5 because of a fluctuation in the resetting voltage, resulting for example from a fluctuation in the manufacturing process. Ordinarily, a power supply line is commonly used for resetting and for the source follower circuit of each pixel, in order to simplify the wiring and to achieve a compacter layout, thereby reducing the pixel size and increasing the aperture ratio. Such configuration is adopted also in the present embodiment, and the resetting voltage is selected not as the actual power source voltage but as a value determined by the threshold value  $V_{th}$  of the resetting MOS transistor.

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20 For explaining the circuit configuration of the solid state image pickup apparatus, an equivalent circuit corresponding to a pixel is shown in Fig. 8. A photodiode 101 accumulates a photo-induced charge which is transferred by a transfer switch 102, and the transferred photo-induced charge is temporarily stored in a floating diffusion area 103, amplified by a source follower MOS transistor 104 and is read out by a

vertical selecting MOS transistor 105 to an output line 106. The floating diffusion area is connected to a power source through a resetting MOS transistor 107, and, in the resetting operation by the application of a voltage of 5 V to the gate thereof, the floating diffusion area is reset to a potential of  $5 V - V_{th}$ .

For example, if the gate voltage of the resetting MOS transistor is 5 V for a power source voltage of 5 V, the resetting voltage becomes  $5 V - V_{th}$ . The threshold voltage of the MOS transistor cannot be made completely free from fluctuation because of fluctuation in the fluctuation of the Si layer or in the thickness of the gate insulation layer, so that the resetting voltage inevitably shows a fluctuation.

Also with an increase in the threshold value  $V_{th}$ , the effective resetting voltage is lowered so that the potential well becomes shallower as shown in  $t2'$  in Fig. 7A. In such state, if the gate voltage and the threshold voltage of the transferring MOS transistor remain constant, there will result charge remaining in the channel of the transferring MOS transistor as shown in  $t2'$ . In order to prevent such charge remaining phenomenon, it is effective to lower the gate voltage of the transferring MOS transistor as shown in  $t2''$ .

Consequently the voltage of the gate 501 of the transferring MOS transistor is controlled by the first potential setting circuit 502, independently from the

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5 voltage of the gate 12 of the resetting MOS transistor.  
Also in order to shift from the state shown in  $t_2'$  in  
Fig. 7A to the state shown in  $t_2''$ , the voltage of the  
gate 501 of the transferring MOS transistor is lowered  
to reduce the potential of the channel 5 thereof,  
thereby transferring the charge in the channel to the  
floating diffusion area.

10 More specifically, in case of the electron  
accumulation type as shown in Fig. 7B, for example when  
a gate-on voltage is 5 V for the resetting MOS  
transistor, a reset voltage is  $3.5 \text{ V} = 5 \text{ V} - V_{th}$  ( $V_{th} =$   
1.5 V) for the floating diffusion area, a saturation  
signal voltage is 1.5 V, a saturation voltage is 2 V  
for the floating diffusion area and a depletion voltage  
15 is 1 V for the photodiode, the gate-on voltage  $V_g$  of  
the transferring MOS transistor is so set as to satisfy  
a relation  $1 + V_{th}' < V_g < 2 + V_{th}'$  (for example  $2.5 \text{ V}$   
 $< V_g < 3.5 \text{ V}$ ). As an example,  $V_g$  is selected as 3.3 V.

20 If the resetting voltage varies from 3.5 V to 3.2  
V by the fluctuation of the process, the saturation  
voltage at the floating diffusion area becomes 1.7 V,  
so that the charge remains in the channel of the  
transferring MOS transistor unless  $V_g$  is so selected as  
to satisfy a condition  $2.5 \text{ V} < V_g < 3.2 \text{ V}$ .

25 In such case, the proper operation without the  
retentive charge can be achieved by changing  $V_g$  from  
3.3 V to for example 3.0 V so as to satisfy the

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above-mentioned condition.

Now there will be given, with reference to Fig. 9, a detailed explanation on the first potential setting circuit shown in Figs. 6A and 6B.

5 Fig. 9 shows a solid state image pickup apparatus 110 for converting incident light into an electrical signal for output to the exterior and a voltage supply unit for driving the solid state image pickup apparatus 110, both apparatus and unit being formed on a same  
10 semiconductor chip.

In the solid state image pickup apparatus 110, there is only shown a pixel, while other pixels, horizontal shift register etc. are omitted.

15 In Fig. 9, the circuit in a broken-lined frame is same as that shown in Fig. 8. In Fig. 9, there are shown a vertical shift register 108 for outputting pulses in succession, and AND circuits 109, 109' and 109" each having two input/output terminals and an electric power source terminal for driving the AND  
20 circuit. In this AND circuit, once high voltage is applied to both two input terminals, a voltage value of the electric power source voltage for driving of the AND circuit is output from the output terminal. As this AND circuit, an AND circuit having a CMOS  
25 structure may be available, for example.

There are also shown a voltage supply unit 113 such as a battery, a voltage conversion circuit 114 for

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converting the voltage of the voltage supply unit 114 into a desired voltage, variable resistors 115, 116, a power supply voltage input terminal 112 for entering a voltage for driving the solid state image pickup apparatus, and an input terminal 111 for entering a voltage for driving the transferring MOS transistor. In addition, there are shown signal lines 105', 102' and 107' for respectively supplying pulses  $\phi_{SEL}$ ,  $\phi_{TX}$  and  $\phi_{RES}$  to be used to turn on and off the vertical selecting MOS transistor 105, the transfer switch 102 and the resetting MOS transistor 107. Under a condition that a pulse is applied from the vertical shift register to one of input terminals of each of the three AND circuits, the AND circuit 109 outputs the voltage value input from the terminal 112, when the pulse  $\phi_{SEL}$  is applied to the signal line 105'; the AND circuit 109' outputs the voltage value input from the terminal 111, when the pulse  $\phi_{TX}$  is applied to the signal line 102'; and the AND circuit 109" outputs the voltage value input from the terminal 112, when the pulse  $\phi_{RES}$  is applied to the signal line 107'. In the present embodiment, the AND circuit 109, the input terminal 111 and the variable resistor 116 constitute the first voltage setting circuit 502 shown in Figs. 6A and 6B.

Thus, even if the threshold voltage  $V_{th}$  of the resetting MOS transistor 107 increases to reduce the

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effective resetting voltage whereby the potential well becomes shallow as shown by  $t_2'$  in Fig. 7A, the above-described configuration can solve the situation in the following manner.

5           As the voltage entered from the input terminal 111 can be made different, by the variable resistor, from the voltage entered from the power supply voltage input terminal 112, the potential well can be realized in a form as shown by  $t_2''$  in Fig. 7A by reducing the voltage entered from the input terminal 111. Thus, in the present embodiment, the pulse voltage for driving the transferring MOS transistor can be adjusted independently from the pulse voltage for driving the resetting MOS transistor. Consequently, even if the  
10           resetting voltage is lowered by the fluctuation in the manufacturing process, the ideal potential well can be formed by adjusting the pulse for driving the  
15           transferring MOS transistor.

Such operation prevents the presence of the photo-induced charge remaining in the transfer channel 5, thereby providing advantages such as:

- 20           a) securing a constant amount of the total transferred charge, thereby ensuring a constant sensitivity; and
- 25           b) avoiding generation of the retentive image.

Fig. 10 is a view showing the potentials of various areas for explaining a second embodiment of the



present invention. In contrast to the first embodiment in which the proper transferring operation is realized, even in the presence of fluctuation in the manufacturing process, by controlling the voltage of the transferring MOS transistor, the present embodiment is to achieve a similar effect by controlling the voltage of the resetting MOS transistor. The circuit configuration corresponding to Fig. 10 is similar to that shown in Figs. 6A and 6B and will not, therefore, be explained further.

In the following there will be explained, with reference to Fig. 11, the second setting circuit shown in Figs. 6A and 6B.

The circuit shown in Fig. 11 is different from that shown in Fig. 9 in that the voltage entered from the input terminal 111 is supplied to the gate of the resetting MOS transistor 107 and the voltage entered from the power supply voltage input terminal 112 is supplied to the gate of the transferring MOS transistor 102. In the present embodiment, the AND circuit 109, the input terminal 111 and the variable resistor 116 constitute the second potential setting circuit 503 shown in Figs. 6A and 6B.

Thus, even if the threshold voltage  $V_{th}$  of the resetting MOS transistor 107 increases to reduce the effective resetting voltage whereby the potential well becomes shallow as shown by  $t_2'$  in Fig. 10, the

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above-described configuration can solve the situation in the following manner.

As the voltage entered from the input terminal 111 can be made different, by the variable resistor, from the voltage entered from the power supply voltage input terminal 112, the potential well can be realized in a form as shown by  $t_2$  in Fig. 10 by elevating the voltage entered from the input terminal 111. Thus, in the present embodiment, the pulse voltage for driving the resetting MOS transistor can be adjusted independently from the pulse voltage for driving the transferring MOS transistor. Consequently, even if the potential of the floating diffusion area comes close to that of the transferring MOS transistor by the fluctuation in the manufacturing process, the ideal potential well can be formed by adjusting the pulse for driving the resetting MOS transistor.

The state at a time  $t_2$  in Fig. 10 is same as that at a time  $t_2$  in Fig. 6B. Also in the present embodiment, the operations at the time  $t_0$  and  $t_1$  are same as those in Fig. 6B.

As shown at a time  $t_2'$  in Fig. 10, the resetting voltage may vary for example because of fluctuation in the manufacturing process, resulting in a charge remaining in the transfer channel 5. In the present embodiment, however, the voltage of the reset gate 12 is independently controlled by the second potential

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109, the input terminal 111' and the variable resistor 116' constitute a second potential setting circuit.

Also the first and second potential setting circuits may be provided, as shown in Fig. 13, in a same semiconductor chip in which the solid state image pickup circuit is formed. Stated differently, the solid state image pickup apparatus may include a voltage adjusting circuit 117 capable of supplying different voltages, for adjusting the voltages supplied to the transferring MOS transistor and the resetting MOS transistor.

The circuit shown in Fig. 13 is so designed as to independently set the power supply voltage for driving the solid state image pickup apparatus, the voltage supplied to the gate of the transferring MOS transistor and the voltage supplied to the gate of the resetting MOS transistor, but the voltage supplied to the gate of the transferring MOS transistor or that supplied to the gate of the resetting MOS transistor may be made common with the power supply voltage. In Fig. 13, the first and second potential setting circuits are constituted by the AND circuit 109 and a voltage adjusting circuit.

As explained in the foregoing first and second embodiments, the present invention allows to avoid remaining of the photo-induced charge in the transfer channel even in the presence of fluctuation of the manufacturing process, thereby providing advantages

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such as:

a) securing a constant amount of the total transferred charge, thus maintaining a constant sensitivity; and

5           b) preventing generation of the retentive image.

Many widely different embodiments of the present invention may be constructed without departing from the spirit and scope of the present invention. It should be understood that the present invention is not limited to the specific embodiments described in the  
10           specification, except as defined in the appended claims.

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